## 1 Classical Linear Code (contd.)

**Definition 1** Let  $\vec{g}$  and  $\vec{e}$  be any codeword and error respectively. The syndrome  $\vec{s}$  of  $\vec{e}$  is defined to be

 $\vec{s} = H(\vec{e}) = H(\vec{g} \oplus \vec{e})$ 

Note that the syndrome of  $\vec{e}$  is independent of the encoded data g, a useful property in the quantum setting. For any  $\vec{e_1}$ ,  $\vec{e_2}$  with syndromes  $\vec{s_1}$ ,  $\vec{s_2}$ , we have

$$\begin{array}{ll} (\vec{s}_1 = \vec{s}_2) & \Longleftrightarrow & H(\vec{e}_1) = H(\vec{e}_2) \\ & \Leftrightarrow & H(\vec{e}_1 - \vec{e}_2) = 0 \\ & \Leftrightarrow & D(\vec{e}_1, \vec{e}_2) > d \end{array}$$

where  $D(\vec{e}_1, \vec{e}_2)$  is the Hamming distance between the two errors. In particular, each element in the set of errors  $\{\vec{e}_i\}$  can be corrected as long as  $\forall i$ 

$$\operatorname{wt}(\vec{e}_i) = D(\vec{e}_i, 0) \leq \left\lfloor \frac{d-1}{2} \right\rfloor$$

**Example 2** Consider the [7, 4, 3] code. Let  $\vec{e}_0 = 0$  and  $\vec{e}_i$  be all zeroes except at the *i*<sup>th</sup> entry. Then

$$H(\vec{e_i}) = the \ i^{th} \ column \ of H$$

In other words, the 3-bit syndrome encodes "which bit has an error" in base 2. The Hamming code above can be generalized to have parameters  $[2^r, 2^r - 1 - r, 3]$ , and the decoding property holds for all of them.

## 2 CSS (Calderbank-Shor-Steane) Codes

Consider two linear codes  $C_B = [n, k_B, d_B]$  and  $C_P = [n, k_P, d_P]$ . Then, we may derive codes to correct for quantum bit flip and phase flip errors by doing the following

- 1. Generate  $M_Z$  from  $H_B$  by replacing 0 with I and 1 with Z.
- 2. Generate  $M_X$  from  $H_P$  by replacing 0 with I and 1 with X.

The rows (tensor product of Pauli matrices) are now called parity check stabilizers  $S_i$  with the property that  $\forall i$ 

$$S_i |\psi\rangle_L = |\psi\rangle_L$$

In addition, for the stabilizer generators to commute, we require

$$\begin{split} C_P^{\perp} \leq C_B & \Longleftrightarrow \quad C_B^{\perp} \leq C_P \\ & \Leftrightarrow \quad H_P G_B^T = 0 \\ & \Leftrightarrow \quad H_B G_P^T = 0. \end{split}$$

**Example 3** 7-bit Steane Code. Both  $C_P$  and  $C_B$  are taken to be the [7, 4, 3] Hamming code.

$$M_{Z} = \begin{pmatrix} Z & Z & Z & I & Z & I & I \\ Z & Z & I & Z & I & Z & I \\ Z & I & Z & Z & I & I & Z \end{pmatrix}$$
$$M_{X} \begin{pmatrix} X & X & X & I & X & I & I \\ X & X & I & X & I & X & I \\ X & I & X & X & I & I & X \end{pmatrix}$$

## 3 Explicit codewords

Let  $X_i, Z_i$  be the rows of  $M_X, M_Z$ . Note that

1. For all  $|\psi\rangle$  we have

$$\prod_{i} (I + X_i) \prod_{j} (I + Z_j) |\psi\rangle \in C$$

2. Let  $l \in C_B$ , then

$$\prod_{j} \left( \frac{I + Z_j}{2} \right) |l\rangle = |l\rangle$$

In order to obtain an explicit characterization of the codewords, we have

$$\begin{split} |l\rangle_L &= \frac{1}{\sqrt{2^{n-k+p}}} \prod_i (I+X_i) \prod_j \left(\frac{I+Z_j}{2}\right) |l\rangle \\ &= \frac{1}{\sqrt{2^{n-k+p}}} \prod_i (I+X_i) |l\rangle \\ &= \frac{1}{|C_p^{\perp}|} \sum_{w \in C_p^{\perp}} |l+w\rangle \end{split}$$

So, there are  $\frac{2^{k_B}}{2^{n-k_P}}$  orthogonal  $|l\rangle_L$ . That gives the correct number of basis states for the stabilizers. As an example, we obtain the following description for the logical encoding of the 7-bit Steane code

$$\begin{split} |0\rangle_l &= \frac{1}{\sqrt{8}} \quad (|0000000\rangle + |1010101\rangle + |0110011\rangle + |1100110\rangle \\ &+ \quad |0001111\rangle + |1011010\rangle + |0111100\rangle + |1101001\rangle) \\ |1\rangle_l &= \frac{1}{\sqrt{8}} \quad (|1111111\rangle + |0101010\rangle + |1001100\rangle + |0011001\rangle \end{split}$$

+ 
$$|1110000\rangle + |0100101\rangle + |1000011\rangle + |0010110\rangle)$$